



# SGM8521/2/4

## 150kHz, 4.7µA, Rail-to-Rail I/O CMOS Operational Amplifiers

### PRODUCT DESCRIPTION

The SGM8521 (single), SGM8522 (dual) and SGM8524 (quad) are rail-to-rail input and output voltage feedback amplifiers offering low cost. They have a wide input common-mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V and the maximum recommended supply voltage is 5.5V. All are specified over the extended -40°C to +125°C temperature range.

The SGM8521/2/4 provide 150kHz bandwidth at a low current consumption of 4.7µA per amplifier. Very low input bias currents of 0.5pA enable the SGM8521/2/4 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-Rail inputs and outputs are useful to designers buffering ASIC in single-supply systems.

Applications for these amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The SGM8521 is available in the tiny SOT23-5 and SO-8 packages. The SGM8522 comes in the miniature SO-8 and MSOP-8 packages. The SGM8524 is offered in TSSOP-14 and SO-14 packages.

### APPLICATIONS

- ASIC Input or Output Amplifier
- Sensor Interface
- Piezo Electric Transducer Amplifier
- Medical Instrumentation
- Mobile Communication
- Audio Output
- Portable Systems
- Smoke Detectors
- Mobile Telephone
- Notebook PC
- PCMCIA Cards
- Battery-Powered Equipment

### FEATURES

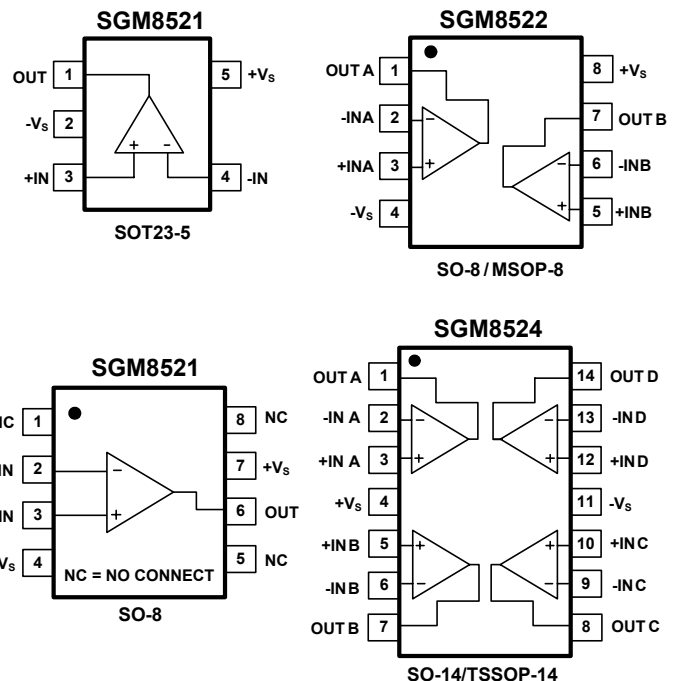
- Low Cost
- Rail-to-Rail Input and Output
- 0.8mV Typical  $V_{OS}$
- Unity Gain Stable
- Gain Bandwidth Product: 150kHz
- Very Low Input Bias Currents : 0.5pA
- Operates on 2.1V to 5.5V Supplies
- Input Voltage Range = -0.1V to +5.6V with  $V_S = 5.5V$
- Low Supply Current: 4.7µA/Amplifier
- Small Packaging

SGM8521 Available in SO-8 and SOT23-5

SGM8522 Available in SO-8 and MSOP-8

SGM8524 Available in SO-14 and TSSOP-14

### PIN CONFIGURATIONS (Top View)



**PACKAGE/ORDERING INFORMATION**

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
SGM8521	SGM8521XN5/TR	SOT23-5	Tape and Reel, 3000	8521
	SGM8521XS/TR	SO-8	Tape and Reel, 2500	SGM8521XS
SGM8522	SGM8522XS/TR	SO-8	Tape and Reel, 2500	SGM8522XS
	SGM8522XMS/TR	MSOP-8	Tape and Reel, 3000	SGM8522XMS
SGM8524	SGM8524XS14/TR	SO-14	Tape and Reel, 2500	SGM8524XS14
	SGM8524XTS14/TR	TSSOP-14	Tape and Reel, 3000	SGM8524XTS14

**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, V+ to V- .....	7.5V	SOT23-5, $\theta_{JA}$ .....	190°C/W
Common-Mode Input Voltage .....	(-V <sub>S</sub> ) - 0.5V to (+V <sub>S</sub> ) + 0.5V	SO-8, $\theta_{JA}$ .....	125°C/W
Storage Temperature Range.....	-65°C to +150°C	MSOP-8, $\theta_{JA}$ .....	216°C/W
Junction Temperature .....	160°C	Lead Temperature Range (Soldering 10 sec).....	260°C
Operating Temperature Range.....	-55°C to +150°C	ESD Susceptibility	
Package Thermal Resistance @ T <sub>A</sub> = 25°C		HBM.....	4000V
		MM.....	400V

**NOTE:**

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**CAUTION**

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

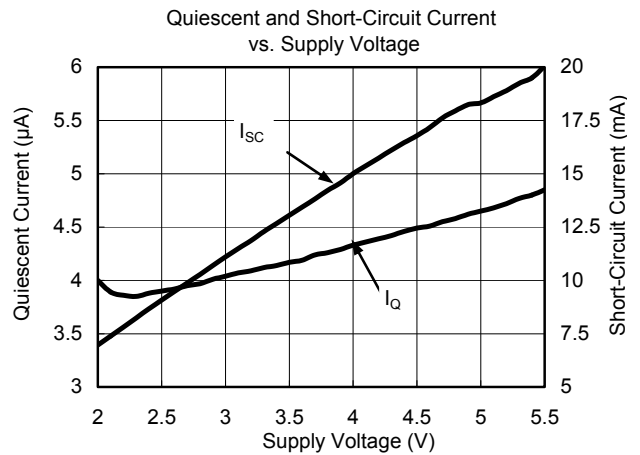
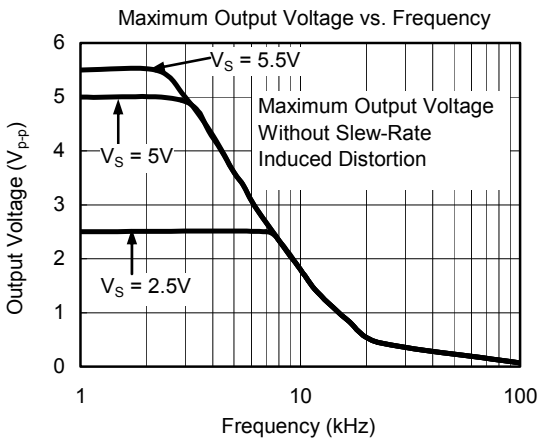
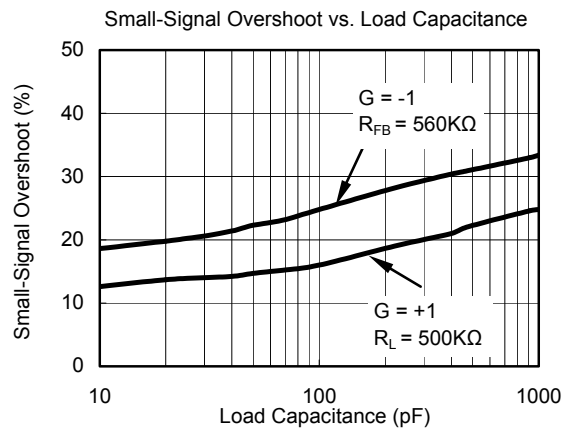
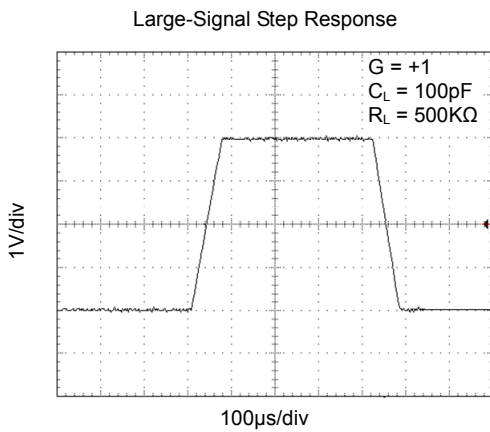
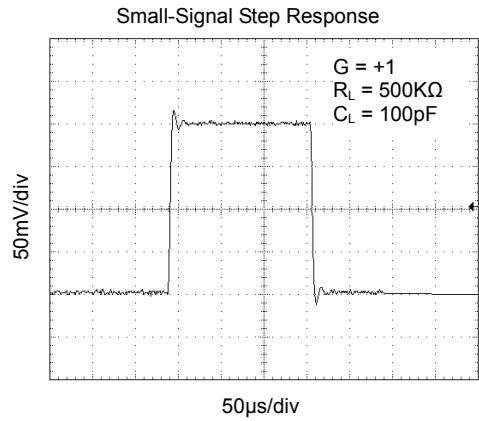
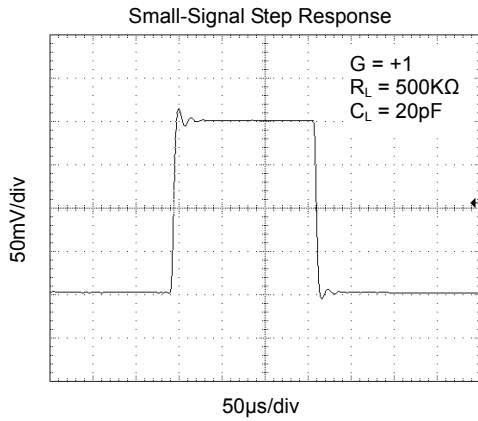
**ELECTRICAL CHARACTERISTICS:  $V_S = +5V$** (At  $R_L = 500K\Omega$  connected to  $V_S/2$  and  $V_{OUT} = V_S/2$ , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	SGM8521/2/4			
			TYP	MIN/MAX OVER TEMPERATURE		
			+25°C	+25°C	UNITS	MIN/ MAX
<b>INPUT CHARACTERISTICS</b>						
Input Offset Voltage	$V_{OS}$		$\pm 0.8$	$\pm 3.5$	mV	MAX
Input Bias Current	$I_B$		0.5		pA	TYP
Input Offset Current	$I_{OS}$		0.5		pA	TYP
Common-Mode Voltage Range	$V_{CM}$	$V_S = 5.5V$	-0.1 to +5.6		V	TYP
Common-Mode Rejection Ratio	CMRR	$V_S = 5.5V, V_{CM} = -0.1V$ to 4V	91	72	dB	MIN
		$V_S = 5.5V, V_{CM} = -0.1V$ to 5.6V	83	63	dB	MIN
Open-Loop Voltage Gain	$A_{OL}$	$R_L = 100K\Omega, V_o = 0.1V$ to 4.9V	100	84	dB	MIN
		$R_L = 500K\Omega, V_o = 0.015V$ to 4.965V	104	90	dB	MIN
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta T$		1.7		$\mu V/^\circ C$	TYP
<b>OUTPUT CHARACTERISTICS</b>						
Output Voltage Swing from Rail		$R_L = 500K\Omega$	0.005		V	MAX
Output Current	$I_{OUT}$		22		mA	MIN
<b>POWER SUPPLY</b>						
Operating Voltage Range				2.1	V	MIN
				5.5	V	MAX
Power Supply Rejection Ratio	PSRR	$V_S = +2.5V$ to + 5.5V $V_{CM} = (-V_S) + 0.5$	91		dB	MIN
Quiescent Current / Amplifier	$I_Q$	$I_{OUT} = 0$	4.7		$\mu A$	MAX
<b>DYNAMIC PERFORMANCE</b>						
Gain-Bandwidth Product	GBP	$C_L = 100pF$	150		kHz	TYP
Slew Rate	SR	$G = +1, 2V$ Output Step	0.04		V/ $\mu s$	TYP
Settling Time to 0.1%	$t_s$	$G = +1, 2V$ Output Step	32		$\mu s$	TYP
<b>NOISE PERFORMANCE</b>						
Voltage Noise Density	$e_n$	$f = 1kHz$	40		$nV/\sqrt{Hz}$	TYP
		$f = 10kHz$	12		$nV/\sqrt{Hz}$	TYP

Specifications subject to change without notice.

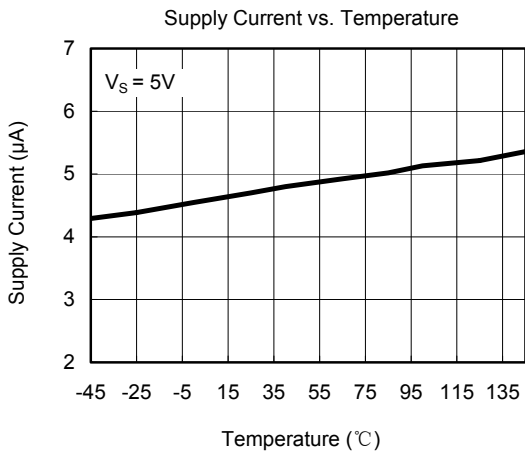
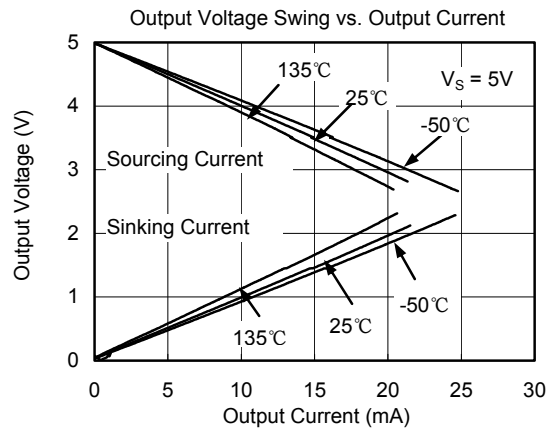
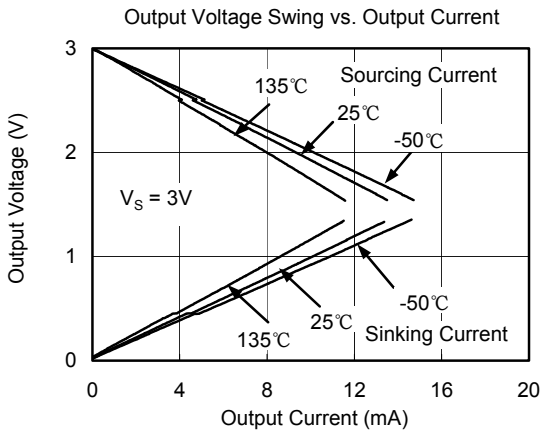
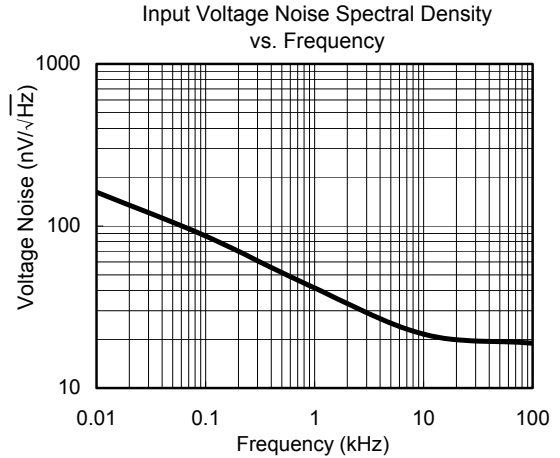
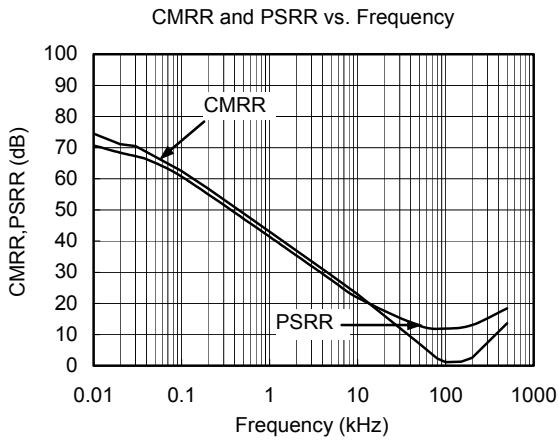
**TYPICAL PERFORMANCE CHARACTERISTICS**

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 500\text{K}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



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APPLICATION NOTES

Driving Capacitive Loads

The SGM852X can directly drive 250pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of amplifiers and this results in ringing or even oscillation. Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor  $R_{ISO}$  and the load capacitor  $C_L$  form a zero to increase stability. The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. Note that this method results in a loss of gain accuracy because  $R_{ISO}$  forms a voltage divider with the  $R_{LOAD}$ .

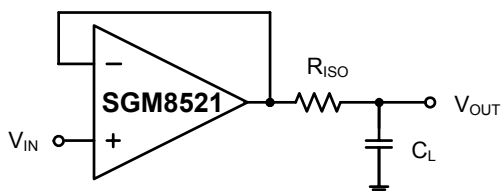


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability.  $R_F$  provides the DC accuracy by connecting the inverting signal with the output,  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

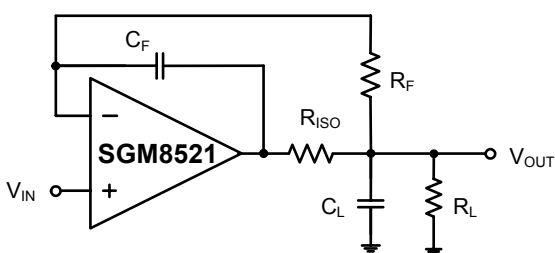


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For no-buffer configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

Power-Supply Bypassing and Layout

The SGM852X family operates from either a single +2.5V to +5.5V supply or dual  $\pm 1.25V$  to  $\pm 2.75V$  supplies. For single-supply operation, bypass the power supply  $V_{DD}$  with a 0.1μF ceramic capacitor which should be placed close to the  $V_{DD}$  pin. For dual-supply operation, both the  $V_{DD}$  and the  $V_{SS}$  supplies should be bypassed to ground with separate 0.1μF ceramic capacitors. 2.2μF tantalum capacitor can be added for better performance.

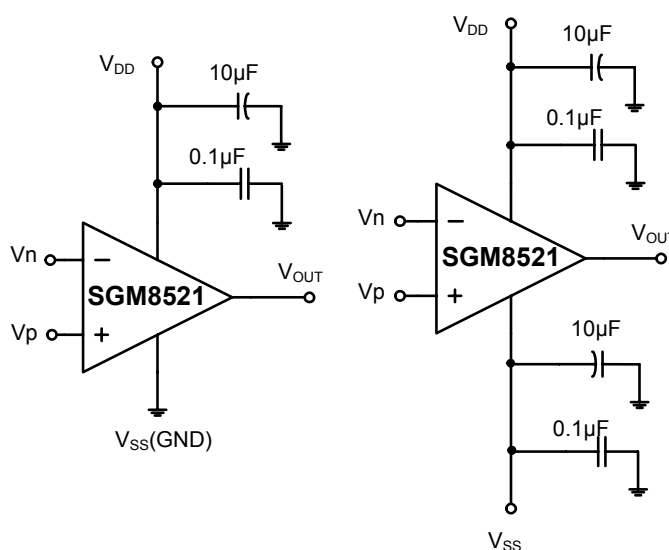


Figure 3. Amplifier with Bypass Capacitors

**TYPICAL APPLICATION CIRCUITS**

**Differential Amplifier**

The circuit shown in Figure 4 performs the difference function. If the resistors ratios are equal ( $R_4 / R_3 = R_2 / R_1$ ), then  $V_{OUT} = (V_p - V_n) \times R_2 / R_1 + V_{REF}$ .

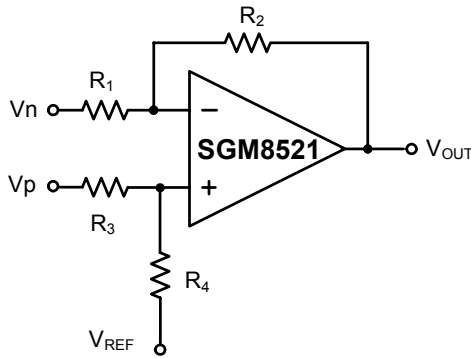


Figure 4. Differential Amplifier

**Instrumentation Amplifier**

The circuit in Figure 5 performs the same function as that in Figure 4 but with the high input impedance.

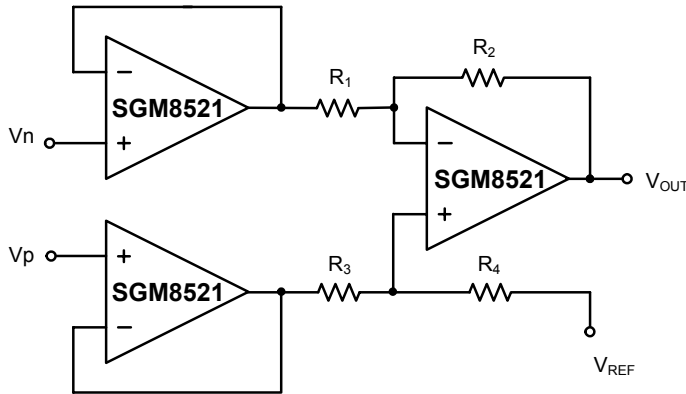


Figure 5. Instrumentation Amplifier

**Low Pass Active Filter**

The low pass filter shown in Figure 6 has a DC gain of  $(- R_2 / R_1)$  and the  $-3\text{dB}$  corner frequency is  $1/2\pi R_2 C$ . Make sure the filter is within the bandwidth of the amplifier. The Large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

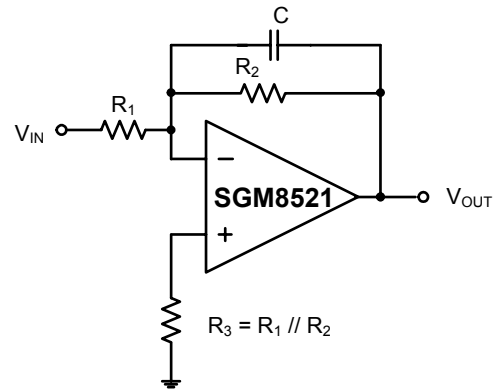
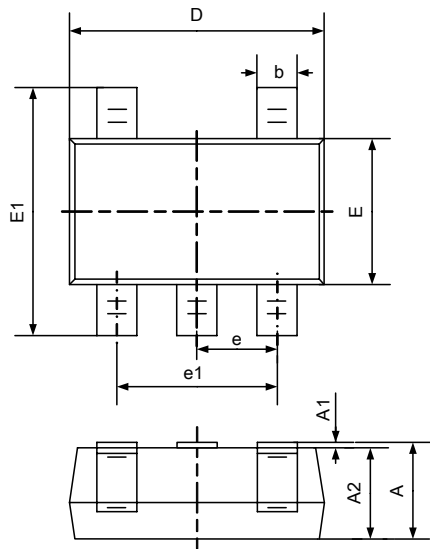


Figure 6. Low Pass Active Filter

PACKAGE OUTLINE DIMENSIONS

SOT23-5

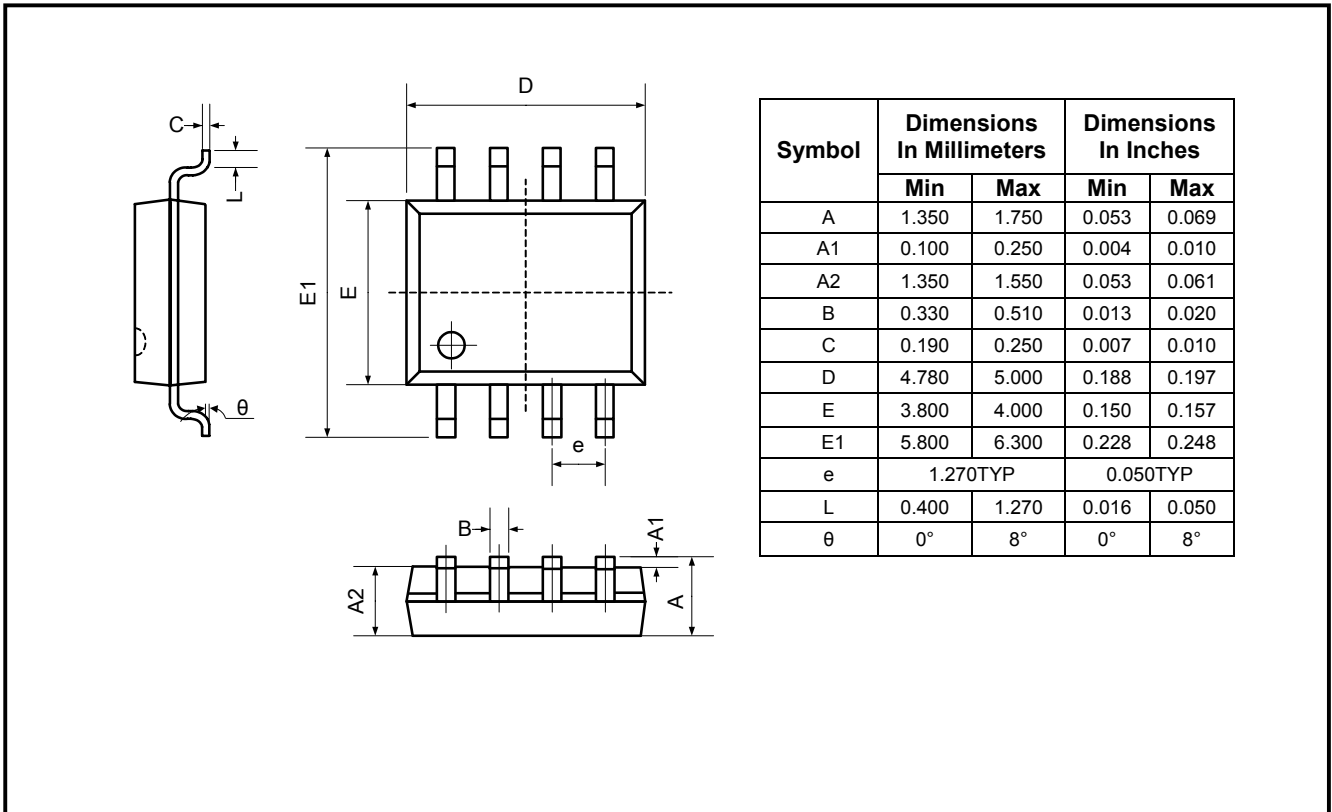


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.400	0.012	0.016
C	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950TYP		0.037TYP	
e1	1.800	2.000	0.071	0.079
L	0.700REF		0.028REF	
L1	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°



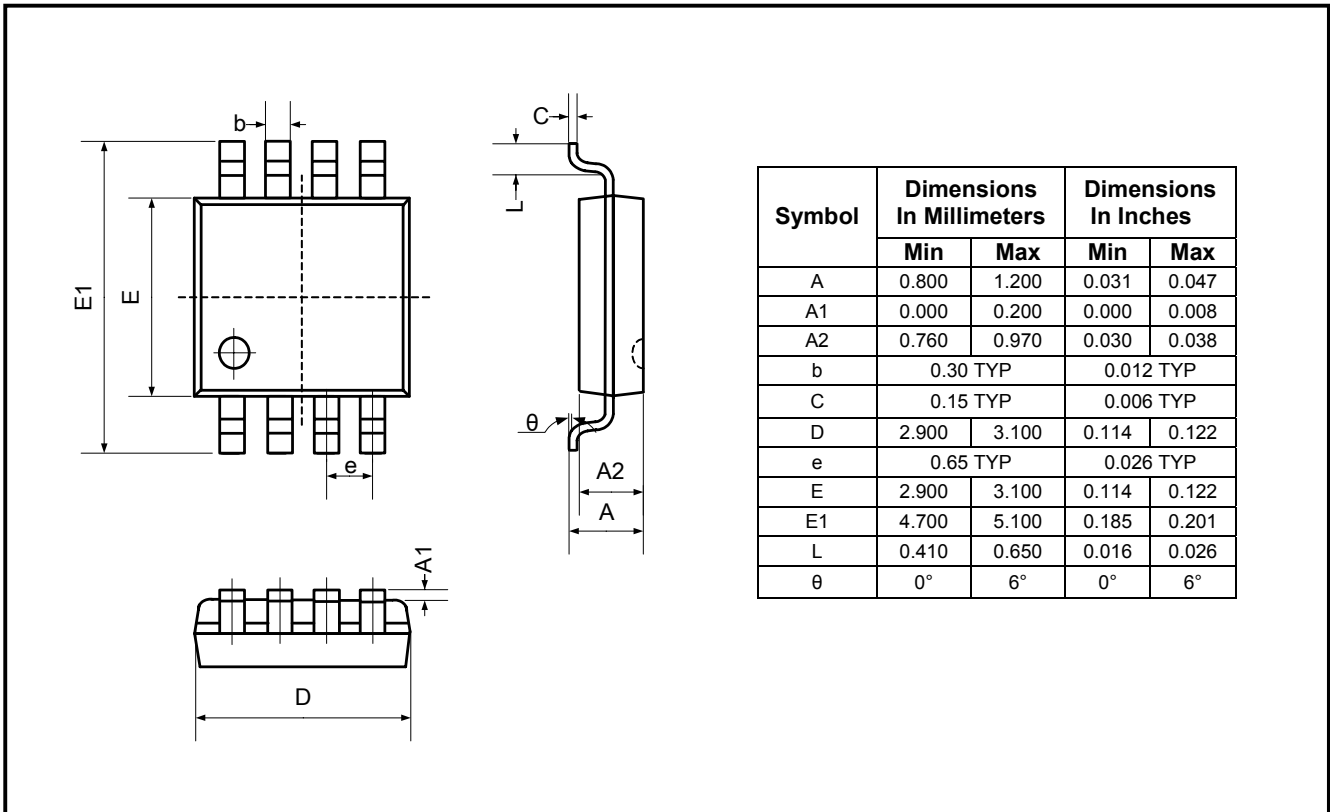
PACKAGE OUTLINE DIMENSIONS

SO-8



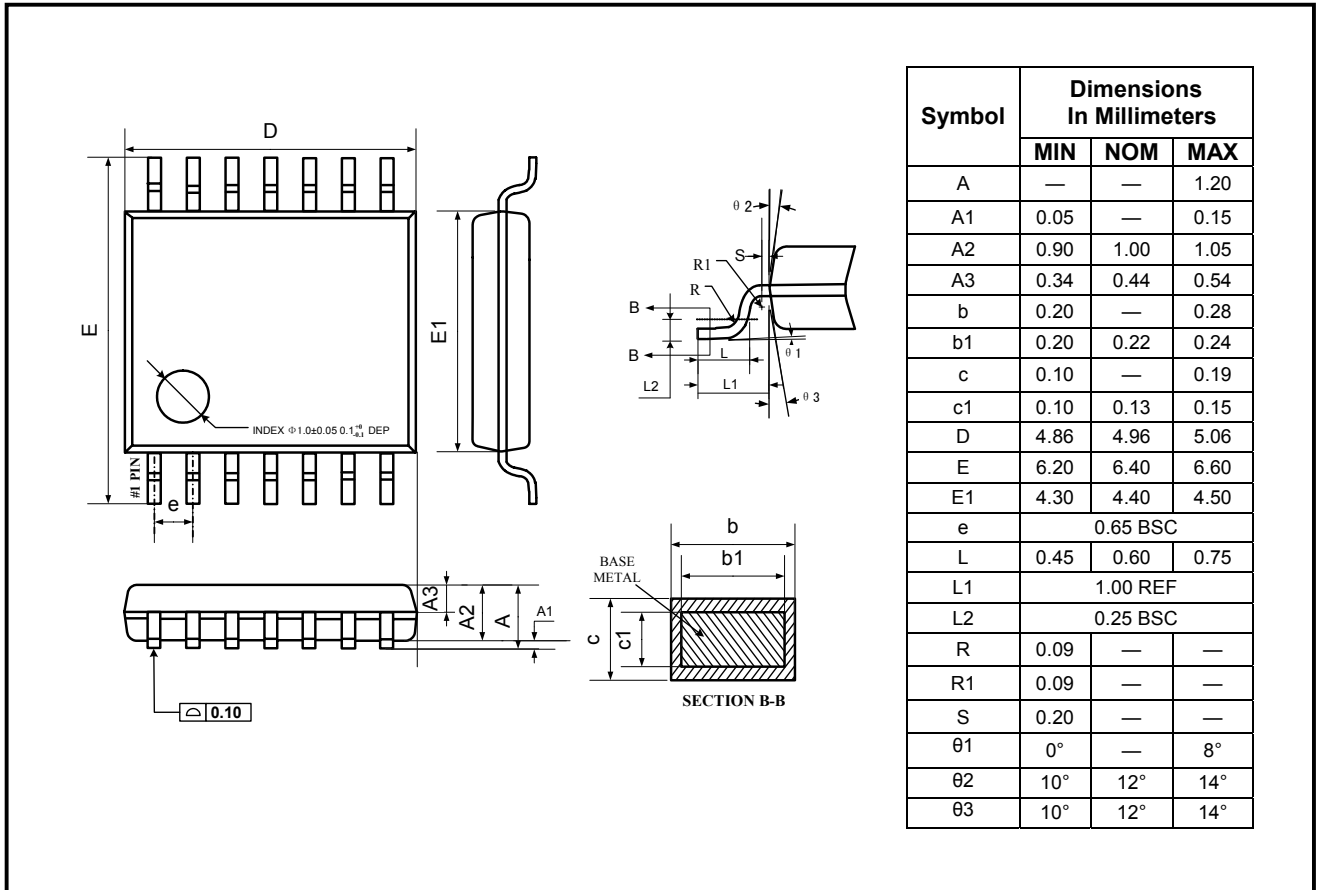
PACKAGE OUTLINE DIMENSIONS

MSOP-8



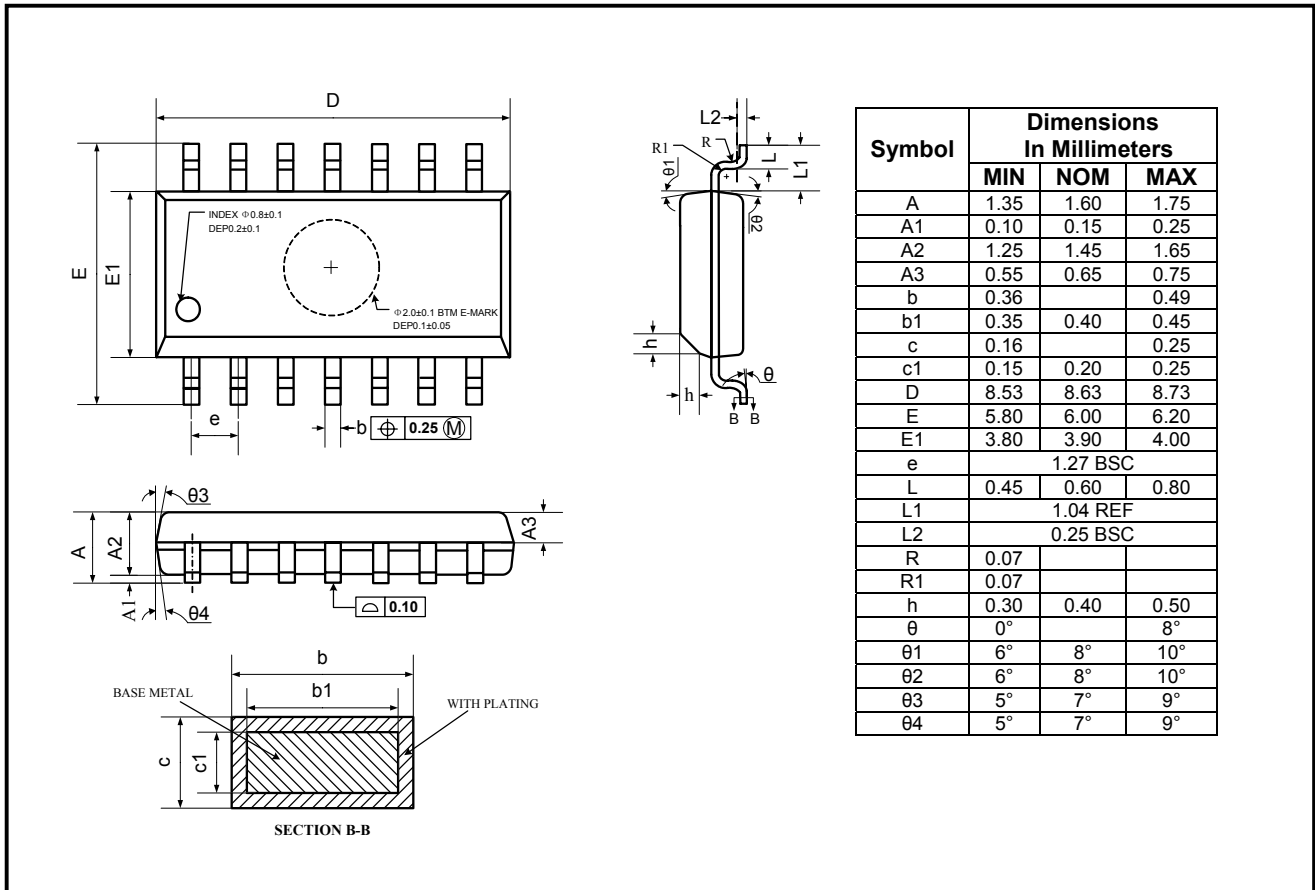
PACKAGE OUTLINE DIMENSIONS

TSSOP-14



PACKAGE OUTLINE DIMENSIONS

SO-14



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